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# CCSDS Unsegmented Code (CUC) & CCSDS Time Manager (CTM)

# Synthesizable VHDL Cores

**Data Sheet** 

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# **1 INTRODUCTION**

This document defines the functionality of the CCSDS Unsegmented Code (CUC) and the CCSDS Time Manager (CTM) synthesizable VHDL cores.

# **1.1** Applicable documents

- AD1 CCSDS 301.0-B-2: Recommendation: Time Code Formats, Blue Book, April 1990, *www.ccsds.org*
- AD2 Packet Telemetry Standard, ESA PSS-04-106, Issue 1, January 1988
- AD3 AMBA<sup>TM</sup> Specification (Rev 2.0), ARM<sup>TM</sup> IHI 0011A, 13th May 1999, Issue A, first release, ARM Limited, *www.arm.com*

# **1.2** Applicable VHDL source code

- AD4 AMBA synthesizable VHDL package, version 0.5, file amba.vhd,
- AD5 CCSDS Unsegmented Code (CUC) VHDL core, version 0.2, file cuc.vhd
- AD6 CCSDS Time Manager (CTM) VHDL core, version 0.2, file ctm.vhd

References AD4 to AD6 can be obtained from www.estec.esa.int/microelectronics

# **1.3** Reference documents

- RD1 CCSDS 102.0-B-4: Recommendation: Packet Telemetry, Blue Book, November 1995, www.ccsds.org
- RD2 ESA VHDL Modelling Guidelines, ASIC/001, Issue 1, September 1994, www.estec.esa.int/microelectronics
- RD3 IEEE Standard VHDL Language Reference Manual, IEEE Std 1076-1993
- RD4 IEEE Standard Multivalue Logic System for VHDL Model Interoperability (Std\_Logic\_1164), IEEE Std 1164-1993
- RD5 IEEE Standards Interpretations: IEEE Standard VHDL Language Reference Manual, IEEE Std 1076/INT-1991

# **1.4** Acronyms and abbreviations

- AHB Advanced High-performance Bus
- APB Advanced Peripheral Bus
- AMBA Advanced Microcontroller Bus Architecture
- ARM Advanced RISC Machines
- CCSDS Consultative Committee for Space Data Systems
- TAI Temps Atomique International
- VHDL VHSIC Hardware Description Language
- VHSIC Very High Speed Integrated Circuits

# 2 FUNCTIONAL DESCRIPTION

# 2.1 Summary of operation

The CCSDS Unsegmented Code (CUC) synthesizable VHDL core provides basic time keeping functions such an Elapsed Time (ET) counter according to the CCSDS Unsegmented Code specification, AD1. It provides support for setting, sampling and correlating the ET counter. It also comprises a Frequency Synthesizer (FS) with which a binary frequency is generated to drive the ET counter. It provides support for setting the increment rate of the ET counter as well as of the FS counter.

The CCSDS Time Manager (CTM) synthesizable VHDL core provides some basic time services based on the ET counter implemented in the embedded CCSDS Unsegmented Code (CUC) synthesizable VHDL core. It provides datation services that sample the ET counter value on external events. It provides alarm services that generate an interrupt when the ET counter value matches arbitrary alarm times. It provides an independent counter that maintains a fine time CUC compliant counter, for the generation of periodic pulses with periods less than one second. This counter is not effected by time setting, sampling or correlation of the aforementioned ET counter. It provides also a dedicated datation service for sampling the ET counter value on the occurrence of the time strobe generated by the Packet TeleMetry Encoder (PTME), generating a Standard Spacecraft Time Source Packet according to the ESA Packet Telemetry Standard, AD2. All services in the CTM VHDL core and those comprised in the embedded CUC VHDL core are accessible via a primary AMBA APB slave interface. All incoming events and outgoing alarms and pulses are also reported through a general interrupt signal that is handled by an interrupt manager. The Standard Spacecraft Time Source Packet is also available through a secondary AMBA APB slave interface on the CTM VHDL core.

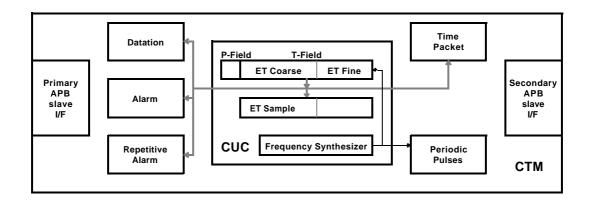


Figure 1: Simplified block diagram of the CUC and CTM cores

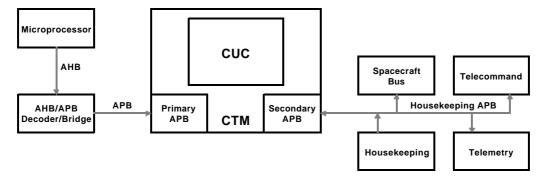
# 2.2 Description of a foreseen system using the VHDL cores

The general approach to accurately maintain onboard time is to have a central time reference measuring the elapsed time from an arbitrary epoch and to distribute regularly this time information to onboard applications by means of messages and synchronisation pulses. Another approach would be to have a centralised time system, where each application that needs to time stamp data could request the unit maintaining the central time reference to provide the relevant time information. Such an approach would have several inherent drawbacks, e.g. in systems with many users, the accuracy of a time stamp could be jeopardised due to long service latency and excessive bus traffic could degrade the overall performance of the data handling system. The purpose CUC and CTM VHDL cores is to provide a building block for such time distribution services by providing the means for CCSDS compliant time keeping and a set of basic user time services.

Most time distribution implementations have required support from the application processor to maintain synchronisation between the central and the local time references. Protocols and formats for distributing time information have differed between spacecraft and have sometimes only provided low resolution or poor accuracy. The purpose of the CTM VHDL core is to provide an accurate time coherence throughout the spacecraft, but it does not implement the time distribution itself since this can be made in several ways depending on e.g. usage of onboard data buses, communication protocol selection etc.

The correlation between the central time reference and ground has already been foreseen by providing a time strobe from the Packet Telemetry encoder (PTME) VHDL core. The time strobe has a deterministic relationship to the bit structure of the telemetry frame. This makes it possible to establish the time relation between the assertion of this time strobe onboard and the reception of the relevant frame on ground, taking into account the down link propagation delay. Each CTM VHDL core instance maintains its own copy of the central elapsed time reference with which onboard applications can time stamp their data. This unbroken chain of time relationships onboard, and between the spacecraft and ground, provides a solution to the problem of knowing when an event took place onboard a spacecraft in any given space-time frame.

The CTM VHDL core is foreseen to fit any type of spacecraft data management systems and instruments due to its generic functionality. The core is foreseen to be used both as a central elapsed time reference in the spacecraft data management system, as well as the local elapsed time reference in an instrument or other subsystem. By using standardised AMBA APB slave interfaces, the integration of the CTM VHDL core should be simple for most systems. The availability of a primary and secondary AMBA APB interface allows the user to connect the CTM VHDL core to both a primary APB controlled by e.g. a microprocessor as well as to a secondary APB controlled by e.g. an autonomous unit generating housekeeping telemetry packet, as shown in figure 2.



**Figure 2:** Block diagram of a system using the CUC and CTM VHDL cores

# 2.3 Functions not included

The CCSDS Time Manager (CTM) synthesizable VHDL core does not implement a protocol for time distribution. It does not perform automatic time synchronisation although accurate time correlation is possible. No time quality issues are addressed.

#### 2.4 Data formats

All Elapsed Time information handled by the CUC and CTM VHDL cores is compliant with the *CCSDS Unsegmented Code* defined in AD1 and repeated hereafter.

# 2.4.1 CCSDS Unsegmented Code: Preamble Field (P-Field)

The time code preamble field (P-Field) may be either *explicitly* or *implicitly* conveyed. If it is implicitly conveyed (not present with T-Field), the code is not self-identified, and identification must be obtained by other means. As presently defined, the explicit representation of the P-Field is limited to one octet whose format is described in table 1.

Bit		Value	Interpretation
0		0	Extension flag
1 - 3	001	1958 January 1 epoch (Level 1) $^{1}$	Time code identification
	010	Agency-defined epoch (Level 2) <sup>2</sup>	
4 - 5	(n	umber of octets of <i>coarse</i> time) - 1	Detail bits for information on the code
6 - 7		(number of octets of <i>fine</i> time)	

**Table 1:** CCSDS Unsegmented Code P-Field definition

- <sup>1</sup> For the 1958 epoch, bits 4 to 5 must be set to "11" to ensure a long enough ambiguity period.
- <sup>2</sup> For the Standard Spacecraft Time Source Packet defined in the ESA Packet Telemetry Standard, bits 1 to 3 must be set to "010". This value is however selectable between "001" and "010" in the CUC VHDL core.

# 2.4.2 CCSDS Unsegmented Code: Time Field (T-Field)

For the unsegmented binary time codes described herein, the T-Field consists of a selected number of contiguous time elements, each element being one octet in length. An element represents the state of 8 consecutive bits of a binary counter, cascaded with the adjacent counters, which rolls over at a modulo of 256, as shown in table 2.

	CCSDS Unsegmented Code												
Preamble		Time Field											
Field		Coarse time						Fine	time				
	2 <sup>31</sup>	$2^{31}$ $2^{24}$ $2^{23}$ $2^{16}$ $2^{15}$ $2^{8}$ $2^{7}$ $2^{0}$						2-1	2-8	2-9	2 <sup>-16</sup>	2-17	2-24

Table 2:	CCSDS Unsegme	ented Code	T-Field definition
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The basic time unit is the second. The T-Field consists of 1 to 4 octets of *coarse* time (seconds) and 0 to 3 octets of *fine* time (sub seconds). The coarse time code elements are a count of the number of seconds elapsed from the epoch. Four octets of coarse time results in a maximum ambiguity period of approximately 136 years.

This allows a time code representation of time through the year 2094 for those which are referenced to the TAI epoch of 1958 January 1. The CCSDS-Recommended epoch is that of 1958 January 1 (TAI), but other Agency-defined epochs may be accommodated as a Level 2 code.

Zero to three octets of *fine* code elements result in a resolution of, respectively: 1 second;  $2^{-8}$  second (about 4 ms);  $2^{-16}$  second (about 15 ms); or  $2^{-24}$  second (about 60 ns).

This time code is not UTC-based and leap second corrections do not apply.

# 2.5 Numbering and naming conventions

Convention according to the CCSDS recommendations, applying to time structures:

- The *most* significant bit of an array is located to the *left*, carrying index number zero.
- An octet comprises eight bits.

	CCSDS n-bit field	
most significant		least significant
0	1 to n-2	n-1

**Table 3:***CCSDS n-bit field definition* 

Convention according to AMBA<sup>TM</sup> Specification, applying to the APB interface:

- Signal names are in upper case, except for the following:
- A lower case 'n' in the name indicates that the signal is active low.
- Constant names are in upper case.
- The *least* significant bit of an array is located to the *right*, carrying index number zero.

	AMBA n-bit field	
most significant		least significant
n-1	n-2 downto 1	0

# **Table 4:**AMBA n-bit field definition

General convention, applying to all other signals and interfaces:

- Signal names are in mixed case.
- An upper case '\_N' suffix in the name indicates that the signal is active low.

# 2.6 CCSDS Unsegmented Code (CUC)

# 2.6.1 General

The CCSDS Unsegmented Code (CUC) synthesizable VHDL core provides basic time keeping functions such an Elapsed Time (ET) counter according to the CCSDS Unsegmented Code specification, AD1. It provides support for setting, sampling and correlating the ET counter. It also comprises a Frequency Synthesizer (FS) with which a binary frequency is generated to drive the ET counter. It provides support for setting the increment rate of the ET counter as well as of the FS counter.

# 2.6.2 Elapsed Time (ET)

The local Elapsed Time (ET) counter is based on a default 32 bit coarse time field and a 24 bit fine time field, complying to the CUC T-Field, AD1. The width of the two time fields can be set by means of generics, determining the number of octets for each field. The counter implementing the ET is incremented on the system clock only when enabled by the frequency synthesizer described below. The ET is incremented with a programmable increment value, which should match the synthesised frequency (section 2.6.3). The width of the increment input can be set by means of a generic, determining the number of bits. The local ET is output in the CUC format, P-Field and T-Field, to be used by an application embedding the CUC VHDL core, e.g. the CTM VHDL core. The P-Field is automatically derived from the generics. The Time Code Identifier can be set to either "001" or "010", being run time selectable.

# 2.6.3 Frequency Synthesizer (FS)

The binary frequency required to determine the ET counter increment is derived from the system clock using a default 24 bit frequency synthesizer. The width of the frequency synthesizer can be set by means of a generic, determining the number of bits. The frequency synthesizer is incremented with a programmable increment value, which should match the available system clock frequency. The output of the frequency synthesizer is used for enabling the increment of the local ET as described above, as well as for driving other counters implemented by applications embedding the CUC core.

# 2.6.4 Time setting and correlation

It is possible to set the local ET counter with an input value. The format of the input is the CUC T-Field. It is possible to reset the phase in the frequency synthesizer. It is possible to sample the local ET counter value. This register is primarily be used for the purpose of time correlation. The implementation of the register can be enabled or disabled by means of a generic. The format of the output is the CUC T-Field. It is possible to correlate an input reference time value with the sampled ET value and the current local ET counter value. This is done by calculating the difference between the input reference time value with the sampled ET value and the current local ET counter value. The implementation of the correlation can be enabled or disabled by means of a generic. The implementation of the correlation can be enabled or disabled by means of a generic. The data format is the CUC T-Field.

#### 2.6.5 Programmable registers, state after reset and error handling

There are not programmable registers. The only registers are the ET and FS counters, and the sample register. All registers are cleared at reset. There is only one operational mode. The core is fully synchronous. There is no explicit error handling. For configuration of the ET and FS counters, see section 2.6.6.

#### 2.6.6 Frequency synthesis and time increment configuration

The increment values for the ET and FS counters depend on the implemented width of each counter and the frequency of the available on the system clock. A procedure for calculating these increment values, the resulting frequency and drift is provided hereafter.

Calculate the sub second resolution corresponding to the highest possible ideal synthesized frequency that can be obtained from the available system clock frequency, being constrained by the implemented width of the fine time part of the ET counter:

$$resolution_{high} = \begin{cases} gFine \cdot 8, \lfloor \ln_2 f_{sys} \rfloor \ge gFine \cdot 8\\ \lfloor \ln_2 f_{sys} \rfloor, \lfloor \ln_2 f_{sys} \rfloor < gFine \cdot 8 \end{cases}$$

Calculate the ideal frequency synthesized from the above sub second resolution:

$$f_{synth_{ideal}} = 2^{resolution_{hig}}$$

Calculate the increment value for frequency synthesis, based on the width of the implemented FS counter, the synthesized ideal frequency and the system clock frequency:

$$IncFrequency = \left[2^{gFrequency} \cdot f_{synth_{ideal}} / f_{sys}\right]$$

Calculate the increment value for the elapsed time, based on the width of the implemented fine time part of the ET counter and the highest obtainable resolution:

$$IncTime = 2^{gFine \cdot 8 - resolution_{high}}$$

Calculate the obtained synthesized frequency from the increment value:

$$f_{synth_{abrained}} = IncFrequency \cdot f_{sys} / 2^{gFrequency}$$

Calculate the static drift caused by the difference between the ideal and obtained synthesized frequency:

static drift = 
$$f_{synth_{ideal}} - f_{synth_{obtained}}$$

Calculate the obtained time resolution:

time resolution = 
$$2^{-resolution_{high}}$$

Calculate the obtained time duration before counter wrap around:

time duration = 
$$2^{gCoarse \cdot 8}$$

# 2.7 CCSDS Time Manager (CTM)

#### 2.7.1 General

The CCSDS Time Manager (CTM) synthesizable VHDL core is compliant with the CCSDS Unsegmented Code format, AD1. It embeds the CCSDS Unsegmented Code (CUC) synthesizable VHDL core.

In addition to the functions embedded in the CUC VHDL core, the CTM VHDL core provides datation, alarm, repetitive alarm, fixed pulses and programmable periodic pulses, as well as generates a Standard Spacecraft Time Source Packet according to AD2. The width and format of all registers are configurable via a subset of generics as available for the CUC VHDL core.

The interfaces to the CTM core are based on two independent AMBA APB slave interfaces, according to AD3.

# 2.7.2 Interfaces

The CTM comprises two independent AMBA APB slave interfaces: a primary interface for general access to all resources and services; and a secondary interface limited to the read-out of the Standard Spacecraft Time Source Packet. The only common signals are the AMBA APB clock and reset. The address inputs of the APB slave interfaces are incompletely decoded. All time services, including the Elapsed Time counter in the embedded CUC VHDL core, are clocked by the AMBA APB clock *PCLK*.

There are four external event input signal that can be used with the datation registers, as described in section 2.7.3. They can also be used with the Standard Spacecraft Time Source Packet register described in section 2.7.8, and for the time setting and correlation, and frequency synthesizer reset as described in section 2.6.

There are two external alarm outputs that are used for the alarm and the repetitive alarm generation described in section 2.7.4 and section 2.7.5. Note that the alarm outputs are asserted one APB clock period after the corresponding increment of the ET counter embedded in the CUC VHDL core.

There is an output that is asserted once every second, as described in section 2.7.6. There is an output that is asserted every time the frequency synthesizer counter wraps around, as described in section 2.7.6. There are four external pulse outputs that are used for the periodic pulse generation described in section 2.7.7.

There is one interrupt output that is used for reporting all incoming events, and outgoing alarms and pulses, which is handled by the interrupt manager described in section 2.7.9.

All input signals are assumed to be synchronous with the AMBA APB interface clock *PCLK*. No input signal synchronisation is performed in the core. All outputs are synchronous with the AMBA APB interface clock, except when explicitly stated in section 3.2.

# 2.7.3 Datation

The CTM VHDL core comprises two datation registers for the purpose of datation of user events relative the ET counter in the embedded CUC VHDL core. These two registers are provided in addition to the ET sample register in the embed CUC VHDL core, which is primarily used for time correlation etc. A fourth datation register is provided for sampling the ET counter when generating a Standard Spacecraft Time Source Packet as described in section 2.7.8.

For all four registers, it is possible to sample the ET counter on an external event or via a register access. For each register it is possible to independently select the source that will trigger the sample. It can either be one of the four external event signals or by register access via the primary AMBA APB slave interface. It is possible to enable or disable the occurrence of a sample by via the aforementioned interface. Each datation service is automatically disabled after an occurrence. The format of all four registers is compliant to the CUC T-Field. The fourth datation register is also accessible via the secondary APB slave interface as described in section 2.7.8.

# 2.7.4 Alarm

The CTM VHDL core comprises one alarm that is generated relative the ET counter in the embedded CUC VHDL core. The alarm is generated on the Alarms(0) output. The time on which the alarm will be generated is programmable via the primary AMBA APB slave interface. It is possible to enable or disable the occurrence of an alarm via the aforementioned interface. The alarm is automatically disabled after an occurrence. The width of the alarm output pulse is one AMBA APB clock period. Note that the alarm output is asserted one APB clock *PCLK* period after the corresponding increment of the ET counter embedded in the CUC VHDL core. The format of the alarm time is compliant to the CUC T-Field.

# 2.7.5 Repetitive alarm

The CTM VHDL core comprises one periodically repeated alarm that is generated relative the ET counter in the embedded CUC VHDL core. The repetitive alarm is generated on the *Alarms(1)* output. The width of the repetitive alarm output pulse is one AMBA APB clock period. Note that the repetitive alarm output is asserted one APB clock *PCLK* period after the corresponding increment of the ET counter embedded in the CUC VHDL core.

The time on which the alarm will be generated is programmable via the primary AMBA APB slave interface. It is possible to mask an arbitrary number of bits when performing comparison between the local ET and the alarm time, resulting in a periodically generated alarm output. It is possible to enable or disable the occurrence of an alarm via the aforementioned interface. The format of the alarm time and the mask is compliant to the CUC T-Field. Bits that are set to logical zero in the mask are excluded in the comparison.

# 2.7.6 Fixed pulses

The CTM VHDL core comprises two fixed period pulses that are propagated from the CUC VHDL core. The *Seconds(0)* output is asserted when the Frequency Synthesizer counter is wrapped around. The *Seconds(1)* output is asserted when the ET counter one second bit position is incremented. The width of the output pulses is one AMBA APB clock *PCLK* period. It is possible to enable or disable individually the occurrence of the pulse outputs via the primary AMBA APB slave interface.

# 2.7.7 Periodic pulses

The CTM VHDL core provides support for generating periodic pulses on four outputs derived from a dedicated CUC fine time counter. The counter can be reset independently from the CUC ET counter, providing a constant frequency. The counter is driven by the Frequency Synthesizer that is embedded in the CUC core. The time increment is the same as for the CUC ET counter. The periodic pulses are generated on the *Pulses(0:3)* outputs. The width of the periodic output pulses is one AMBA APB clock *PCLK* period. The frequencies on which the pulses will be generated are programmable as masks via the primary AMBA APB slave interface. The format of the mask is compliant to the fine time part of the CUC T-Field. Bits that are set to logical zero in the mask are excluded in the comparison with the counter all zero value. It is possible to enable or disable individually the occurrence of the periodic pulse outputs via the aforementioned interface.

# 2.7.8 Standard Spacecraft Time Source Packet

The CTM VHDL core comprises one datation register for sampling the ET counter when generating a Standard Spacecraft Time Source Packet according to the ESA Packet Telemetry Standard, AD2. It is possible to sample the ET counter on an external event or via a register access. It is possible to select the source that will trigger the sample. It can either be one of the four external event signals, or by register access via the primary AMBA APB slave interface. It is possible to enable or disable the occurrence of a sample via the aforementioned interface. This datation register can be accessed via the primary APB slave interface and the format is compliant to the CUC T-Field. This datation register can also be accessed via the secondary APB slave interface. The format then complies to either the CUC T-Field or to the Standard Spacecraft Time Source Packet format described in AD2. The selection is made by means of a generic. In case the latter is selected, a Source Sequence Count is implemented as well, see table 10.

# 2.7.9 Interrupt manager

The CTM VHDL core comprises an interrupt manager that combines the discrete alarm and pulse outputs described above, and the discrete event inputs into a single interrupt output *Interrupt*. Each interrupt source can be individually masked, cleared and set. The interrupt output is asserted until all pending interrupt sources have been cleared. All accesses to the interrupt manager are via the primary AMBA APB slave interface.

#### 2.7.10 Programmable registers and operational modes

The primary CTM AMBA APB slave interface supports 32 bit wide data input and output. The input address is interpreted as a byte address, as per AD4. Since each access is a word access, the two least significant address bits are assumed always to be zero, only address bits 6:2 are decoded. Misaligned addressing is not supported. For read accesses, data output is produced combinatorially from address etc., un-mapped bits are always driven to zero, is designed to work in a multiplexed unidirectional bus scheme. Re-mapping between the opposing numbering conventions in the CCSDS and AMBA documentation is performed. When the CCSDS field is narrower than the AMBA data width, zeros are padded to the left. The primary AMBA APB slave interface provides direct access to the P-Field and T-Field of the ET counter embedded in the CUC VHDL core. It should be noted that the T-Field counter value is not frozen during read out and that it can change between reading the coarse and the fine time registers.

Register name	Address	Read/Write	Remarks
Configuration Register	00h	R/W	See table6
Service Register	04h	R/W	See table7
Increment Frequency Register	08h	R/W	See section3.1.1.3
Increment Time Register	0Ch	R/W	See section3.1.1.4
Set/Correlate Time Coarse Register	10h	R/W	T-Field, coarse part
Set/Correlate Time Fine Register	14h	R/W	T-Field, fine part
Sample Time Coarse Register	18h	R	T-Field, coarse part
Sample Time Fine Register	1Ch	R	T-Field, fine part
Datation Coarse 0 Register	20h	R	T-Field, coarse part
Datation Fine 0 Register	24h	R	T-Field, fine part
Datation Coarse 1 Register	28h	R	T-Field, coarse part
Datation Fine 1 Register	2Ch	R	T-Field, fine part
Alarm Coarse Register	30h	R/W	T-Field, coarse part
Alarm Fine Register	34h	R/W	T-Field, fine part
Repetitive Alarm Coarse Register	38h	R/W	T-Field, coarse part
Repetitive Alarm Fine Register	3Ch	R/W	T-Field, fine part
Repetitive Alarm Coarse Mask Register	40h	R/W	T-Field, coarse part
Repetitive Alarm Fine Mask Register	44h	R/W	T-Field, fine part
Periodic Pulse Fine 0 Mask Register	48h	R/W	T-Field, fine part
Periodic Pulse Fine 1 Mask Register	4Ch	R/W	T-Field, fine part
Periodic Pulse Fine 2 Mask Register	50h	R/W	T-Field, fine part
Periodic Pulse Fine 3 Mask Register	54h	R/W	T-Field, fine part
Time Packet Coarse Register	58h	R	T-Field, coarse part
Time Packet Fine Register	5Ch	R	T-Field, fine part
Interrupt Manager Register	60h	R/W	See table10
CUC P-Field Register	64h	R	See table1
CUC T-Field Coarse Register	68h	R	T-Field, coarse part
CUC T-Field Fine Register	6Ch	R	T-Field, fine part
Unused addresses	70h:7Ch	R	All zeros

<b>Table 5:</b> CTM VHDL core registers on primary AMBA APB slave intergram	gisters on primary AMBA APB slave interface
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Bit number	Default Name		Remarks
31:12	all zeros	unused	All zeros
11:9	000	Reset Frequency Source	Select event source (see table8)
8:6	000	Set Time Source	Select event source (see table8)
5:3	000	Sample Time Source	Select event source (see table8)
2:0	000	Correlate Time Source	Select event source (see table8)

**Table 6:**Configuration Register bit definition (address 00h) (R/W)

Bit number	Default	Name	Remarks
31:25	all zeros	unused	All zeros
24:22	000	Datation 0 Source (0 to 2)	Select event source (see table8)
21:19	000	Datation 1 Source (0 to 2)	Select event source (see table8)
18	0	Alarm On	Enable alarm output Alarms(0)
17	0	Repetitive Alarm On	Enable repetitive alarm output Alarms(1)
16	0	Frequency On	Enable frequency pulse output Seconds(0)
15	0	Seconds On	Enable one-second pulse output Seconds(1)
14:11	0000	Pulse On (0 to 3)	Enable periodic pulse outputs Pulses(0 to 3)
10:8	000	Pulse Reset Source (0 to 2)	Select event source (see table8)
7:5	000	Time Packet Source (0 to 2)	Select event source (see table8)
4	0	TAI Time Code	TAI compliant time code id when set (see table 1)
3:0	0000	Sample Rate (0 to 3)	Set time sampling rate (see table11)

**Table 7:**Service Register bit definition (address 04h) (R/W)

Bit 0 to 2	Source	Bit 0 to 2	Source
000	Disabled	100	Events(0) input
001	Force Event	101	Events(1) input
010	Disabled	110	Events(2) input
011	Disabled	111	Events(3) input

**Table 8:**Source selection for datation and time adjustment trigger events

Bit number	Default	Name	Remarks
31:30	all zeros	unused	All zeros
29:20	all zeros	Clear	Interrupt source cleared when bit is set (write only)
19:10	all zeros	Mask	Interrupt source enabled when bit is set (read/write)
9:0	all zeros	Source	Pending interrupt source during read Set interrupt source when bit is set during write
-	-	Bit assignment for above	Events(0 to 3) & Alarms(0 to 1) & Pulses(0 to 3)

**Table 9:**Interrupt Manager Register bit definition (address 60h) (R/W)

The secondary CTM AMBA APB slave interface supports 8 bit wide data output. The input address is interpreted as a byte address, as per AD4. Since each access is a word access, the two least significant address bits are assumed always to be zero, only address bits 5:2 are decoded. Misaligned addressing is not supported. For read accesses, data output is produced combinatorially from address etc, is designed to work in a multiplexed unidirectional bus scheme. Only data bits 7:0 are used, the rest are always driven to all zeros. Re-mapping between the opposing numbering conventions in the CCSDS and AMBA documentation is performed. When the CCSDS field is narrower than the AMBA data width, zeros are padded to the left.

Register content	Address	Read/Write	Remarks
Packet Header Octet 0	00h	R	0000000
Packet Header Octet 1	04h	R	0000000
Segment Flags & Sequence Count (0 to 5)	08h	R	00 & 14 bit counter $^1$
Sequence Count (6 to 13)	0Ch	R	
Packet Length (0 to 7)	10h	R	gCoarse + gFine + 1 <sup>1</sup>
Packet Length (8 to 15)	14h	R	
Data Field & Sample Rate (0 to 3)	18h	R	0000 & sampling rate (see table 11) $^{1}$
P-Field	1Ch	R	See table1
T-Field (0 to 7)	20h	R	Always present
T-Field (8 to 15)	24h	R	Only if gCoarse + gFine > 1
T-Field (16 to 23)	28h	R	Only if gCoarse + gFine > 2
T-Field (24 to 31)	2Ch	R	Only if gCoarse + gFine > 3
T-Field (32 to 39)	30h	R	Only if gCoarse + gFine > 4
T-Field (40 to 47)	34h	R	Only if gCoarse + gFine > 5
T-Field (48 to 55)	38h	R	Only if gCoarse + gFine > 6
Unused addresses	3Ch	R	0000000

**Table 10:** CTM VHDL core registers on secondary AMBA APB slave interface

<sup>1</sup> The implementation of registers within the address rages 00h to 18h and 24h to 38h is optional, being selectable by means of generics. The read data value will be all zeros if a register that is not implemented is accessed.

Bit 0 to 3	Rate (in frames)	Bit 0 to 3	Rate (in frames)
0000	1	0101	32
0001	2	0110	64
0010	4	0111	128
0011	8	1000	256
0100	16	others	undefined

**Table 11:**Sampling rate according to AD2

# 2.7.11 Initialisation, state after reset and error handling

All registers are cleared at reset, except the Sequence Count that is set to all ones. There is no explicit error handling. The CTM and CUC cores are synchronous.

#### **3** INTERFACE DESCRIPTION

#### 3.1 CCSDS Unsegmented Code (CUC) interfaces

#### **3.1.1** Synthesis configuration

#### 3.1.1.1 gCoarse: Number of coarse time octets Natural range 1 to 4

This generic selects the number of *octets* to be implemented for the coarse time of the ET.

#### **3.1.1.2** gFine: Number of fine time octets: Natural range 0 to 3

This generic selects the number of *octets* to be implemented for the fine time of the ET.

#### 3.1.1.3 gFrequency: Width of frequency synthesizer: Positive

This generic selects the number of *bits* to be implemented for the frequency synthesizer. It defines the array width in bits for the frequency increment value input.

#### 3.1.1.4 gIncrement: Width of time increment: Positive

This generic defines the array width in bits for the time increment value input. Should be equal to or less than (gCoarse + gFine) \* 8 and 32, depending on the smallest value.

#### 3.1.1.5 gSample: Time sample register support: Natural range 0 to 1

This generic selects whether a time sample register is to be implemented or not.

#### **3.1.1.6** gCorrelate: Time correlation support: Natural range 0 to 1

This generic selects whether a time correlation is to be implemented or not. This requires that the time sample register is implemented.

#### **3.1.2** System interface

#### 3.1.2.1 Reset\_rise\_N: Synchronised reset: Std\_ULogic (I)

This active low input signal asynchronously resets the CUC VHDL core. The signal is assumed to be synchronous with the system clock *Clk* rising edge. The input is used on registers that are all clocked on the rising *Clk* edge.

#### 3.1.2.2 *Clk*: System clock: Std\_ULogic (I)

This input signal is the system clock signal for the CUC VHDL core. All registers are clocked on the rising Clk edge.

# **3.1.3** Configuration interface

# 3.1.3.1 TimeCodeTAI: TAI Time Code Identifier: Std\_ULogic (I)

This input signal selects the Time Code Identifier in the P-Field to be compliant with the TAI format when set, i.e. "001", else it is "010" which is compliant to AD2. The input is passed through combinatorial logic before being sampled on rising Clk edge.

# 3.1.3.2 *IncFrequency*: Increment frequency: UnSigned(0 to gFrequency-1) (I)

This input signal sets the increment of the Frequency Synthesizer counter. The most significant bit is to the left. The input is passed through combinatorial logic before being sampled on rising *Clk* edge.

# 3.1.3.3 IncTime: Increment time: UnSigned(0 to gIncrement-1) (I)

This input signal sets the increment of the Elapsed Time counter. The most significant bit is to the left. The input is passed through combinatorial logic before being sampled on rising Clk edge.

# **3.1.4** Unused outputs (to guide synthesis tools during optimization)

# 3.1.4.1 *FreqAdder*: UnSigned(0 to gFrequency) (I/O)

Frequency Synthesizer counter increment adder. This is a combinatorial output.

# **3.1.4.2** *TimeAdder*: UnSigned(0 to (gCoarse+gFine)\*8-1) (I/O)

Elapsed Time counter increment adder. This is a combinatorial output.

**3.1.4.3** *TimeCorr*: UnSigned(0 to (gCoarse+gFine)\*8-1) (I/O)

Elapsed Time correlation adder. This is a combinatorial output.

# **3.1.5** Elapsed Time counter represented as CUC

# **3.1.5.1** *P\_Field*: CUC P-Field: UnSigned(0 to 7) (O)

This output signal carries the value of the P-Field. Data is output on rising *Clk* edge.

**3.1.5.2** T\_Field: CUC T-Field: UnSigned(0 to (gCoarse+gFine)\*8-1)) (O)

This output signal carries the value of the ET counter T-Field. The most significant bit is to the left. Data is output on rising *Clk* edge.

# **3.1.6** Time set interface

#### 3.1.6.1 *ResetFreq*: Reset frequency: Std\_ULogic (I)

This input signal synchronously resets the Frequency Synthesizer counter when asserted. The reset of the Frequency Synthesizer corresponds to resetting the phase of the Elapsed Time counter. The input is sampled on rising *Clk* edge.

#### 3.1.6.2 *SetTime*: Set time: Std\_ULogic (I)

This input signal synchronously sets the Elapsed Time counter with the *TimeInput* value when asserted. The input is sampled on rising *Clk* edge.

#### 3.1.6.3 SampleTime: Sample time: Std\_ULogic (I)

This input signal synchronously samples the Elapsed Time counter value into the time sample register when asserted. The input is sampled on rising *Clk* edge.

# 3.1.6.4 CorrelateTime: Correlate time: Std\_ULogic (I)

This input signal synchronously correlates the Elapsed Time counter value that is held in the time sample register with that value input on the TimeInput signal, and stores the result back into the ET counter. The input is sampled on rising *Clk* edge.

#### **3.1.6.5** *TimeInput*: Time input: UnSigned(0 to (gCoarse+gFine)\*8-1) (I)

This input signal is used for setting and correlating the Elapsed Time counter. The most significant bit is to the left. The input is sampled on rising Clk edge.

#### **3.1.6.6** *TimeOutput*: Time output: UnSigned(0 to (gCoarse+gFine)\*8-1) (O)

This output signal carries the value of the time sample register. The most significant bit is to the left. Data is output on rising *Clk* edge.

# 3.1.6.7 FreqPulse: Frequency wrap: Std\_ULogic (O)

This output signal is asserted for one *Clk* period when the Frequency Synthesizer counter is wrapped around. The signal is output on rising *Clk* edge.

#### **3.1.6.8** SecondPulse: One second period: Std\_ULogic (O)

This output signal is asserted for one *Clk* period when the ET counter one second bit position is incremented. The signal is output on rising *Clk* edge.

# **3.1.7 Diagnostic support**

# 3.1.7.1 Frequency: System frequency: Natural (I)

This input signal is the system frequency that is used for diagnostics only.

# **3.2** CCSDS Time Manager (CTM) interfaces

#### 3.2.1 Synthesis configuration

#### 3.2.1.1 gCoarse: Number of coarse time octets Natural range 1 to 4

This generic selects the number of *octets* to be implemented for the coarse time of the ET.

#### **3.2.1.2** gFine: Number of fine time octets: Natural range 1 to 3

This generic selects the number of *octets* to be implemented for the fine time of the ET.

# **3.2.1.3** gFrequency: Width of frequency synthesizer: Natural range 1 to 32

This generic selects the number of *bits* to be implemented for the frequency synthesizer.

#### **3.2.1.4** *gIncrement*: Width of time increment: Natural range 1 to 32

This generic defines the array width in bits for the time increment value input. Should be equal to or less than (gCoarse + gFine) \* 8 and 32, depending on the smallest value.

#### **3.2.1.5** gTimeSourcePkt: Time Source Packet support: Natural range 0 to 1

This generic selects whether the Standard Spacecraft Time Source Packet support is to be implemented or not. It effects the secondary AMBA APB slave interface only.

# 3.2.2 Diagnostic support

# 3.2.2.1 Frequency: System frequency: Natural (I)

This input signal is the system frequency that is used for diagnostics only in the CUC.

# **3.2.3** AMBA APB slave interface: system signals

For detailed information on the two first signals see AD3 and AD4.

# 3.2.3.1 PRESETn: Synchronised reset: Std\_ULogic (I)

This active low input signal asynchronously resets the CTM VHDL core and the embedded CUC VHDL core. The signal is assumed to be synchronous with the AMBA APB clock *PCLK* rising edge. The input is used on registers that are all clocked on the rising *PCLK* edge.

# 3.2.3.2 PCLK: Interface clock: Std\_ULogic (I)

This input signal is the AMBA APB clock which is the clock for the CTM VHDL core and the embedded CUC VHDL core. All registers are clocked on the rising *PCLK* edge.

# 3.2.4 Primary AMBA APB slave interface

For detailed information on the records used for the APB interface see AD3 and AD4.

# **3.2.4.1** *APBIn*: Interface input: APB\_Slv\_In\_Type (I)

This signal record is the general APB slave interface input.

# 3.2.4.1.1 *PSEL*: Slave select: Std\_ULogic (I)

This signal indicates that the APB slave device is selected and a data transfer is required. The input is sampled on the rising *PCLK* edge for write accesses.

# 3.2.4.1.2 PENABLE: Enable strobe: Std\_ULogic (I)

This strobe signal is used to time all accesses on the peripheral bus. The enable signal is used to indicate the second cycle of an APB transfer. The rising edge of *PENABLE* occurs in the middle of the APB transfer. The input is sampled on the rising *PCLK* edge for write accesses.

# 3.2.4.1.3 PADDR: Address bus: Std\_Logic\_Vector(PAMAX-1 downto 0) (I)

This is the APB address bus can be up to 32 bits wide. Only address bits 6:2 are decoded. The input is sampled on the rising *PCLK* edge for write accesses. *PAMAX* is defined in the AMBA VHDL package, AD4.

# 3.2.4.1.4 *PWRITE*: Write strobe: Std\_ULogic (I)

This signal indicates the APB transfer direction When asserted this signal indicates an APB write access and when de-asserted a read access. The input is sampled on the rising *PCLK* edge for write accesses.

# 3.2.4.1.5 PWDATA: Write data bus: Std\_Logic\_Vector(PDMAX-1 downto 0) (I)

The APB write data bus is driven by the peripheral bus master during write cycles (when *PWRITE* is asserted). The write data bus can be up to 32-bits wide. The data is sampled on the rising *PCLK* edge for write accesses. *PDMAX* is defined in the AMBA VHDL package, AD4.

# **3.2.4.2** *APBOut*: Interface output: APB\_Slv\_Out\_Type (**O**)

This signal record is the general APB slave interface output.

# 3.2.4.2.1 PRDATA: Read data bus: Std\_Logic\_Vector(PDMAX-1 downto 0) (O)

The APB read data bus is driven by the selected slave during read cycles (when *PWRITE* is de-asserted). The read data bus can be up to 32-bits wide. The data output is produced combinatorially from the address bus etc. *PDMAX* is defined in the AMBA VHDL package, AD4.

# 3.2.5 Secondary AMBA APB slave interface

For detailed information on the records used for the APB interface see AD3 and AD4.

# **3.2.5.1** *APBPktIn*: Interface input: APB\_Slv\_In\_Type (I)

This signal record is the general APB slave interface input. For details on record members see section 3.2.4.1, with the additions listed below.

# 3.2.5.1.1 PADDR: Address bus: Std\_Logic\_Vector(PAMAX-1 downto 0) (I)

Only address bits 5:2 are decoded.

# 3.2.5.2 *APBPktOut*: Interface output: APB\_Slv\_Out\_Type (O)

This signal record is the general APB slave interface output. For details on record members see section 3.2.4.2, with the additions listed below.

3.2.5.2.1 PRDATA: Read data bus: Std\_Logic\_Vector(PDMAX-1 downto 0) (O)

Only data bits 7:0 are used, the rest are always driven to all zeros.

# 3.2.6 Time events

# **3.2.6.1** *Events*: Event input: Std\_Logic\_Vector(0 to 3) (I)

These inputs signal external events to which setting, sampling and correlation of time and general datation can be coupled. The inputs are sampled on the rising *PCLK* edge.

# 3.2.6.2 Alarms: Alarm and repetitive alarm output: Std\_Logic\_Vector(0 to 1) (O)

These outputs signal the occurrence of an alarm. The generate pulse has the width of one *PCLK* period. The outputs are driven on the rising *PCLK* edge.

# 3.2.6.3 Seconds: Fixed periodic pulse output: Std\_ULogic (O)

These output signals carry fixed periodic pulses. The generate pulse has the width of one *PCLK* period. The outputs are driven on the rising *PCLK* edge.

# **3.2.6.4** *Pulses*: Periodic pulse output: Std\_Logic\_Vector(0 to 3) (**O**)

These output signals carry periodic pulses. The generate pulse has the width of one *PCLK* period. The outputs are driven on the rising *PCLK* edge.

# 3.2.6.5 *Interrupt*: General interrupt output: Std\_ULogic (O)

This output signal is the general interrupt signal from the CTM VHDL core. It is asserted until all pending sources are cleared. The output is driven on the rising *PCLK* edge.

## 4 VHDL SOURCE CODE DESCRIPTION

The CUC and CTM Synthesizable VHDL cores and test benches are written according to RD2 as far as applicable. The VHDL code complies to VHDL'93, RD3.

#### 4.1 Packages and libraries, interface port and generic types

The following VHDL packages are used in the CUC VHDL core:

- Std.Standard, Std.TextIO
- IEEE.Std\_Logic\_1164, IEEE.Std\_Logic\_Arith, IEEE.Std\_Logic\_TextIO

The CUC VHDL core interfaces do not comply to the normally recommended Std\_ULogic and Std\_Logic\_Vector types, RD5, since the UnSigned type is used for all arrays. The generics used are not of type Integer, normally supported by synthesis tools, but are based on subtypes of Integer such as Natural and Positive.

The following VHDL packages are used in the CTM VHDL core:

- Std.Standard
- IEEE.Std\_Logic\_1164, IEEE.Std\_Logic\_Arith
- AMBA\_Lib.AMBA

The CTM VHDL core interfaces do comply to the normally required, Std\_ULogic and Std\_Logic\_Vector types, RD2. The generics used are not of type Integer, normally supported by synthesis tools, but are based on subtypes of Integer such as Natural and Positive.

The recommended target library for the CUC and CTM VHDL cores is CUC\_Lib. Note that the they can be located in another library than CUC\_Lib, but this will require a modification of one line in the CTM VHDL code.

Note that the AMBA interface can be located in another library than AMBA\_Lib, but this will also require a modification of one line in the CTM VHDL code.

# 4.2 Special considerations

The CUC VHDL core provides diagnostic support by means of writing information to the standard output when one of the input settings is modified. The information is derived from the diagnostic input signal specifying the real time frequency of the system clock used by the core. The provided information comprises CUC P-Field reporting and frequency and time accuracy reporting. The minimum width of the ET counter is asserted when the Time Code Identifier is set to TAI compliance.

The CUC VHDL core interface has three input/output signals that are used for guiding some synthesis tools during optimization. These signals are all related to different adders in the core and should remain unconnected.

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